

Frequency Synthesiser (I²C BUS Programmable)

Advance Information

DS2429 - 3.2 September 1994

The NJ88C33 is a synthesiser circuit fabricated on Mitel Semiconductor's 1.4 micron CMOS process, assuring very high performance. It is I^2C compatible and can also be programmed at up to 5MHz. It contains a 16-bit R counter, a 12-bit N counter and a 7-bit A counter.

A digital phase comparator gives improved loop stability with current source outputs to reduce loop components. A voltage doubler is provided for the loop driver to improve control voltage range to the VCO when operating at low supply voltages.

FEATURES

- Easy to Use
- Low Power Consumption (15mW)
- Single Supply 2.5V to 5.5V
- Digital Phase Comparator with Current Source Outputs
- Serial (I²C Compatible) Programming, 5MHz max
- Channel Loading in 8µs
- 150MHz Input Frequency Without Prescaler at 4.5V (52MHz at 2.7V)
- Standby Modes
- Use of Two-Modulus Prescaler is Possible

APPLICATIONS

- Cordless Telephones (CT2, DECT)
- Cellular Telephones (GSM, PCN, ETACS)
- Hand Held Marine Radios
- Sonarbuoys
- Video Clock generators

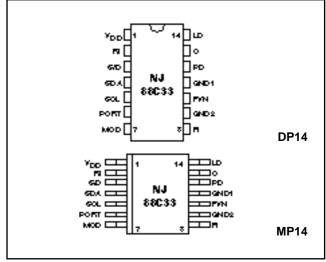


Fig.1 Pin connections (not to scale) - top views

ABSOLUTE MAXIMUM RATINGS

ORDERING INFORMATION

NJ88C33 MA DP (Industrial - Plastic DIL package)
NJ88C33 MA MP (Industrial - Miniature Plastic DIL package)

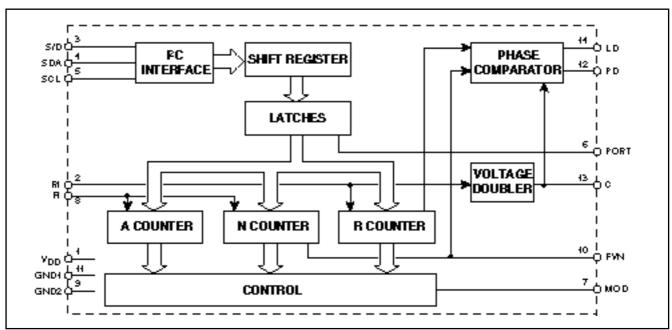


Fig.2 Simplified block diagram of NJ88C33

PIN DESIGNATIONS

Pin No.	Pin Name	Description							
1	V _{DD}	Supply voltage (normally 5V or 3V).							
2	RI	Reference frequency input from an accurate source, normally a crystal oscillator. The input is normally an AC coupled sinewave but may be a DC coupled square wave.							
3	S/D	Single/dual modulus operating mode selection input. Single modulus operation is selected by driving the pin low. 'High' selects dual modulus mode.							
4	SDA	I ² C bus data input pin. It is also an open-drain output for generating I ² C bus acknowledge pulses.							
5	SCL	I ² C bus clock input. It can be clocked at up to 5MHz.							
6	PORT	Output control pin, which can be programmed via the I ² C bus. It can be connected to the S/D pin to select single or dual modulus mode under bus control.							
7	MOD	Modulus control pin. It is high in single modulus mode but switches in dual modulus operation. In dual modulus mode, MOD remains low during operation of the A counter until A=0; MOD then remains high until N=0, when both counters are reloaded. It can be programmed via the I ² C bus as an open-drain or push-pull output.							
8	FI	Frequency input from a VCO or prescaler. The input is normally an AC coupled sinewave but may be a DC coupled square wave.							
9	GND2	Dedicated ground for the FI input buffer. It should be connected to the VCO ground or the prescaler ground, if used. Any noise on this pin will affect the performance of the VCO loop.							
10	FVN	Open-drain output from the N counter.							
11	GND1	Ground supply pin (global).							
12	PD	Tristate current output from the phase detector. The polarity of the output can be programmed via the $\mbox{I}^2\mbox{C}$ bus.							
13	С	Voltage doubler output. The operation of the doubler can be controlled via the I2C bus. In applications where the voltage doubler is switched off, this pin should be connected to GND1; a reservior capacitor should be connected from this pin to GND1 for applications where it is switched on.							
14	LD	Open-drain lock detect output - requires integration if used.							

OPERATING RANGE

Test conditions (unless otherwise stated):

PLL locked, RI = 10MHz

Characteristic	Symbol	Value			Unit	Conditions
Characteristic	Symbol	Min.	Тур.	Max.	Ollit	Conditions
Supply voltage Ambient temperature Supply current	${\sf V}_{\sf DD} \ {\sf T}_{\sf amb}$	2.5 -40	5	5.5 +85	°C V	
Single modulus	l _{DD}		2.1	3.0	mA	FI = 50MHz, V_{FI} = 150mVrms, N,R > 1000 without voltage doubler, V_{DD} = 5V, T_{amb} = 25°C
Dual modulus	l _{DD}		2	3.0	mA	FI = 10MHz, V_{FI} = 500mVrms, N,R > 1000 without voltage doubler, V_{DD} = 5V, T_{amb} = 25°C
Standby mode	I _{DD}			1	μΑ	FI = 50MHz, V_{FI} = 150mVrms, preamp off, divider off, V_{DD} = 5V, T_{amb} = 25°C
Standby mode	I _{DD}		1.0	1.5	mA	FI = 50MHz, V_{FI} = 150mVrms, preamp on, divider off, V_{DD} = 5V, T_{amb} = 25°C

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Fia.3 Functional block diagram
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ELECTRICAL CHARACTERISTICS

These characteristics are guaranteed over the following conditions (unless otherwise stated):

 $V_{DD} = 4.5 V$ to 5.5V, $T_{amb} = -40 ^{\circ} C$ to +85°C INPUT SIGNALS

Characteristic	Symbol	Value			Unit	Conditions
Characteristic	Symbol	Min.	Тур.	Max.	Onic	Conditions
Input Signals SDA, SCL, S/D Input voltage high Input voltage low Input capacitance Input current	V	0.7V _{DD}		V _{DD} 0.3V _{DD} 10	V V pF µA	V _{IN} = V _{DD} = 5.5V
Input signal RI Input frequency Input voltage Input capacitance Input current	f max V Irms C I	100		52 10 10	MHz mV pF µA	Sinewave input Note 1, 2 $V_{IN} = V_{DD} = 5.5V$
Input signal FI Input frequency Input voltage Input capacitance Input current	f _{max} V _{Irms} C _I	50		52 10 10	MHz mV pF µA	Dual modulus operation Sinewave input Note 1, 2 $V_{IN} = V_{DD} = 5.5V$
Input signal FI Input frequency Input voltage Input capacitance Input current	fmax VIrms VIrms VIrms CIIMS	30 100 200		150 10 10	MHz mV mV mV pF µA	Single modulus operation Sinewave input FI = 0-70MHz Note 1, 2 FI = 70-120MHz Note 1, 2 FI = 120-150MHz Note 1, 2 $V_{IN} = V_{DD} = 5.5V$

Note.1Lowest noise floor achieved at 10dB above this level with I2C bus operating. The source impedance should be less than 2k .

Note.2DC coupled input amplitude $V_{IRMS} > 0.8V_{DD}$.

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Oh ann at aniatis	Value				l lmit	O I'II'	
Characteristic	Symbol	Min. Typ.		Max.	Unit	Conditions	
Output Signals SDA, LD Output voltage low	V _{oL}			0.4	V	Open drain, I _{OL} = 3mA	
Output Signal PD High current mode (see Fig.4) Low current mode Tristate	HU HD HD LU LD LD Z	1.9 -1.9 0.475 -0.475	2.5 -2.5 0.625 -0.625 50	3.1 -3.1 0.775 -0.775	mA mA mA mA nA	$\begin{array}{l} C_L = 400 pF, \ tristate \ output \\ 0 < V_{PD} < 4.5, \ V_{DD} = 5V, \ T = 25^{\circ}C \ Note \ 1 \\ 0.4 < V_{PD} < 5, \ V_{DD} = 5V, \ T = 25^{\circ}C \ Note \ 1 \\ 0 < V_{PD} < 4.6, \ V_{DD} = 5V, \ T = 25^{\circ}C \ Note \ 1 \\ 0.4 < V_{PD} < 5, \ V_{DD} = 5V, \ T = 25^{\circ}C \ Note \ 1 \\ T_{amb} = -25^{\circ}C \ to +60^{\circ}C \end{array}$	
Output Signal FVN Output voltage low Output low pulse width	V _{OL} t _{WL}			0.4 1/FI	V	Open drain output I _{OL} = 1mA C _L = 30pF	
Output Signals MOD, PORT Output voltage high Output voltage low	V _{OH}	V _{DD} -0.4		0.4	V V	Push-pull output IOH = 0.5mA I _{OL} = 0.5mA	
Output Signal LD Output voltage low Output low pulse width	V _{OL} t _{WL}	V _{OL}			V ns	Open drain output I _{OL} = 3mA, C _L = 30pF Loop locked Loop not locked FVN = FI/N f _C = RI/R	

Note.1Temperature coefficient for current is typically -0.7%/°C

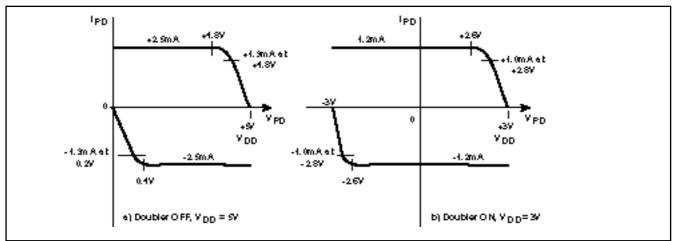


Fig. 4 Typical output signal PD, high current mode

VOLTAGE DOUBLER

V	3V
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Characteristic	Symbol		Value	•	Unit	One ditions	
Characteristic	Syllibol	Min.	Тур.	Max.	Onit	Conditions	
Output Pin C Output voltage	V _c V _c	-V _{DD}		-V _{DD} + 0.8V -V _{DD} + 1.5V	V	$f_{VD} = 2MHz, I_{OC} = 0\mu A, V_{DD} = 3V$ $f_{VD} = 2MHz, I_{OC} = 100\mu A, V_{DD} = 3V$	
Current Consumption	I _D			100	μA	$f_{VD} = 2MHz, I_{OC} = 0\mu A, V_{DD} = 3V$	

TIMING INFORMATION

Characteristic	Symbol		Value		Unit	O and distance			
Characteristic	Syllibol	Min.	Тур.	Max.	Onit	Conditions			
Input Signal RI Input frequency Input frequency Rise time Fall time Slew rate	f _{max} f _{max} t _R t _F	0 0		52 10 1.5 1.5	MHz MHz µs µs V/µs	$V_{DD} = 2.7V$			
Input Signal FI Input frequency Input frequency Rise time Fall time Slew rate	f _{max} f _{max} t _R t _F	0 0 3		52 20 1.5 1.5	MHz MHz µs µs V/µs	Dual modulus $V_{DD} = 2.7V$			
Input Signal FI Input frequency Input frequency Rise time Fall time Slew rate	f _{max} f _{max} t _R t _F	0 0 3		150 52 1.5 1.5	MHz MHz µs µs V/µs	Single modulus $V_{DD} = 2.7V$			
Output Signal PORT Rise time Fall time	t _R t _F			1	μs μs	$C_L = 30pF$ $C_L = 30pF$			
Output Signal FVN Fall time	t _F		20		ns	C _L = 30pF			
Output Signal MOD Rise time Fall time Delay time (L H) Delay time (H L)	t _R t _F t _{DLH} t _{DHL}			10 10 15 15	ns ns ns ns	$C_L = 30 pF$ $C_L = 30 pF$ $C_L = 30 pF$ Measured from +Ve edge of FI $C_L = 30 pF$ Measured from +Ve edge of FI			

PHASE COMPARATOR

The phase comparator produces current pulses of duration equal to the difference in phase between the comparison frequency (fc=Rl/R), and f_{VN} , the divided-down VCO frequency (Fl/N).

When status bit 4 is set high the positive polarity mode of the output PD is selected. When fc leads $f_{_{VN}}$ the PD output goes high; when $f_{_{VN}}$ leads fc it goes low. Similarly, selecting the negative polarity mode of PD by programming bit 4 of the status register low causes PD to have the inverse polarity. The loop filter integrates the current pulses to produce a voltage drive to the VCO.

No pulses are produced when locked. The lock detect output, LD, produces a logic '0' pulse equal to the phase difference between $f_{_{\rm C}}$ and $f_{_{\rm VN}}.$

When the phase difference between fc and $f_{\rm VN}$ is too small to be resolved by the phase detector then no current pulses are produced. In this region the loop does not reduce the close-in noise on the VCO output. This can be overcome using a very high value resistor to leak a few nanoAmps of current from the filter and keep the loop on the edge of the region.

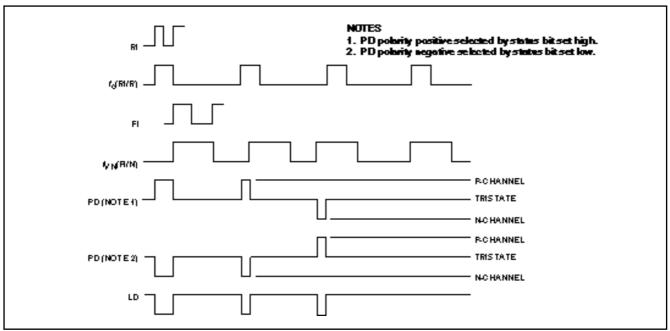


Fig. 5 Phase comparator phase diagram

PROGRAMMING

Transmission Protocol

I²C programming messages consist of an address byte followed by a sub-address byte followed by 1, 2 or 3 bytes of data. Bit 7 of the address byte must match the setting of the S/D pin for the address to be recognised. This allows for separate addressing of two NJ88C33 synthesisers on the same bus. The sub-address should be set to select the correct registers to be programmed and should be followed by the appropriate number of data bytes. Registers are not programmed until the complete message protocol has been checked.

Each message should commence with a START condition and end with a STOP condition unless followed immediately by another transfer, when the STOP condition may be omitted.

Data is transferred from the shift register to the latches on a STOP condition or by a second START condition.

A START condition is indicated by a falling edge on the Serial Data line, SDA, when the Serial Clock line, SCL, is high.

A rising edge on SDA when SCL is high indicates a STOP condition as shown in Fig.6.

Data on SDA is clocked into the NJ88C33 on the rising edge of SCL. The NJ88C33 acknowledges each byte transferred to it by pulling the SDA line low for one cycle of SCL after the last bit has been received.

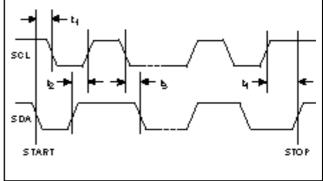


Fig. 6 I²C timing diagram

12C TIMING INFORMATION

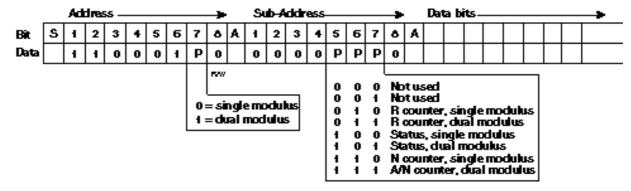
VDD = 4.5V to 5.5V, Tamb = -40°C to +85°C

Parameter	Symbol	Va	Unit	
ranamotor	Oymbo.	Min.	Max.	J
Serial clock frequency	f _{SCL}		5	MHz
SCL hold after START	t ₁	200		ns
Data set-up time	t ₂	20		ns
Data hold after SCL low	t ₃	0		ns
SCL set-up before STOP	t ₄	20		ns

Address and Sub-Address Formats

The correct addressing sequence for the NJ88C33 is shown below. The START condition is followed by the address byte, the acknowledge from the NJ88C33, the sub-

address byte, another acknowledge then the associated data. The correct values for each address and sub-address are listed, together with the message selection options.



Data Formats

Each of the data formats should be preceded contiguously by the addressing sequence given above.

R counter: single or dual modulus

	4								R	Data	· —						-		
Bit	1	2	3	4	5	6	7	٥	A	1	2	3	4	5	6	7	٥	A	St
Data	P	P	P	P	P	P	P	P		P	P	P	P	P	P	P	P		
	MSE								_	•							Les	_	

Status: single or dual modulus

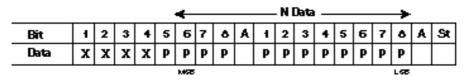
	Data									
Bit	1	2	3	4	5	6	7	٥	A	St
Data	P	P	P	P	P	P	P	P		

	Status Byte												
Bit	0	1											
1	PORT = low	PORT = high											
2	Counters off (1)	Counters on											
3	FI and RI off (2)	FI and RI on											
4	PD = polarity negative	PD = polarity positive											
5	PD bias = 0.625mA	PD bias = 2.5mA											
6	$f_{VD} = RI/2$	$f_{VD} = RI/4$											
7	Doubler off	Doubler on (3)											
8	MOD = push-pull	MOD = open drain											

NOTES

- 1. In this standby mode the counters are disabled but the voltage doubler and I^2C interface can both function.
- 2. In this standby mode the FI and RI preamplifiers are disabled, which stops the counters and the voltage doubler. The I²C interface still operates.
- 3. The voltage doubler should only be used when $V_{\scriptscriptstyle DD}$ 3.0V

N counter: single modulus



A/N counters : dual modulus

						4			A	Data	—		-	+					N D	ata	_					-		
Bit	1	2	3	4	5	6	7	٥	A	1	2	3	4	5	6	7	٥	A	1	2	3	4	5	6	7	٥	A	St
Data	х	х	х	х	P	P	P	P		P	P	P	P	P	P	P	P		P	P	P	P	P	P	P	P		
MSS							LGS MGS									ræ												

APPLICATION CIRCUITS

Single Modulus

In this mode, the NJ88C33 synthesiser can be used with or without a fixed modulus prescaler. The R counter is programmed with a value to produce a comparison frequency fc. When the N counter is changed by 1 the loop is no longer in lock and the phase detector output produces current pulses to bring the loop back into lock. These pulses are integrated by the loop fiiter to produce the VCO voltage drive. When the VCO loop is locked, Fl/N=f_c i.e., the VCO frequency is N x f_c.

Using a prescaler with a division ratio P, the smallest VCO output frequency step is Pf_c and the VCO frequency is PNf_c.

If a low pass filter is connected to the lock detect output as shown and sampled by the microprocessor, the proximity of the synthesiser loop to lock can be evaluated.

The A counter is not used in this mode.

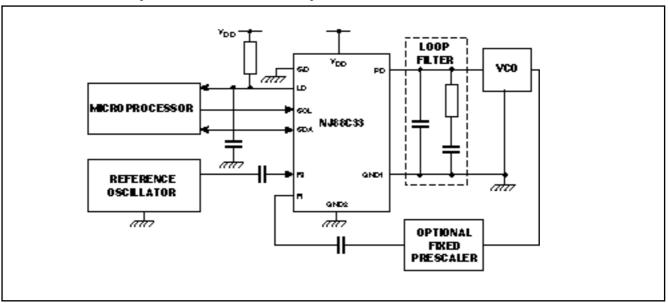


Fig. 7 Single modulus application

Dual Modulus

This mode allows much higher frequencies to be used in conjunction with a prescaler but maintains the step size, fc. In this mode, a dual modulus prescaler (with ratios P and P + 1) must be used with the NJ88C33. The A counter controls the MOD output, which is used to select the division ratio of the prescaler.

When the A counter is non-zero, the MOD output is low and goes high when the A counter has counted down to zero. MOD remains high until the N counter reaches zero, when both counters are re-loaded. Thus, the prescaler divides by P for N-A cycles and by P + 1 for A cycles of FI. The VCO frequency is given by $PNf_C + Af_C$.

Note that programming A = 0 produces a count of 128 cycles.

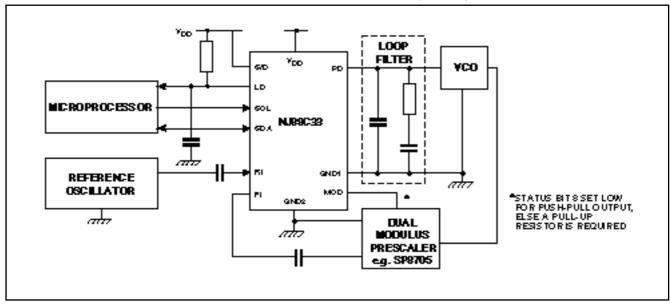


Fig. 8 Dual modulus application

VCO Driving Without Voltage Doubler

To switch off the voltage doubler, bit 7 of the status register is programmed low. This will reduce current consumption and minimise noise. The voltage doubler output C should be connected to GND1 as connection to GND2 would induce noise in the VCO loop.

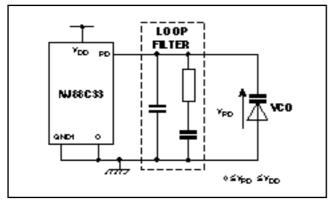


Fig. 9 Driving a VCO without voltage doubler

Further Applications Information

A stand-alone programmer card and an evaluation board are available for evaluating the NJ88C33. The programmer card allows two sets of variables to be programmed into both the divider and status registers during alternate programming cycles, at either the standard I²C bus rate of 100kHz or at 2MHz.

Initialisation is with either a manual push-button or by an external logic level pulse; a synchronisation output is provided to allow a quick assessment of 'step' and 'settle' responses to be made.

VCO Driving With Voltage Doubler

The voltage doubler is switched on by setting bit 7 of the status register high. It is recommended that a reservoir capacitor of at least $1\mu F$ be connected from C to GND1.

The voltage doubler is designed to boost VCO drive in low voltage applications.

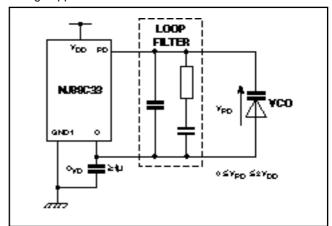


Fig. 10 Driving a VCO using the voltage doubler

The NJ88C33 evaluation board (Fig. 11) dernonstrates the preferred layout technique - providing a reference oscillator, a 60 to 80MHz VCO and a simple loop filter to complete a minimal frequency synthesiser loop. The two units allow analysis of different loop variables as well as the selection of comparison frequencies for fast frequency-hopping loops.

Application Note: AN94, 'Using the NJ88C33 PLL Synthesiser' explains the design equations and demonstrates the use of the device, and is available from your local Mitel Semiconductor customer service centre.

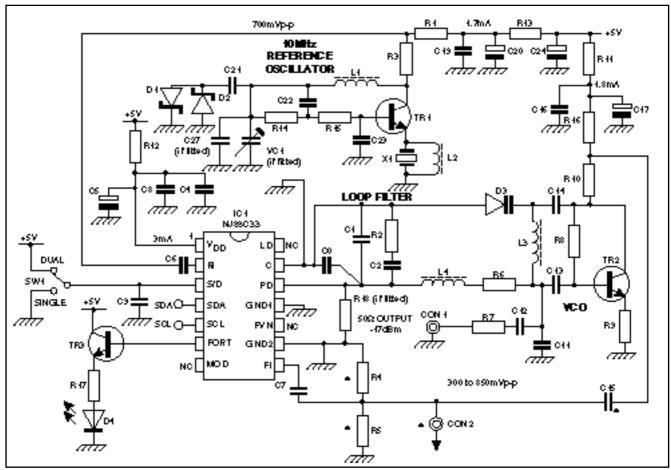


Fig. 11 Typical applications circuit

COMPONENT LIST FOR FIG. 11

COMPONENT LIST FOR FIG. 11												
	Сар	acito	's	Re	sistors		Inductors	Miscellaneous				
C0 C1 C2 C4 C5 C6 C7 C8 C9 C11 C12	1nF 10% 100nF 10% 1µF Tant. 10nF 10% 22µF/35V Elect. 10nF 10% 1nF 10% 1nF 10% 1nF 10% 150pF 5% NPO 1nF 10%	r	22μF/35V Elect. 10nF 10% 22μF/35V Elect. 10nF 10% 10nF 10% 22pF 5% NPO 22μF/35V Elect. 22pF 5% NPO	R1 R2 R3 R4 R5 R6	270 470 330 100 100 1k 120 27k Link 1k	L1 L2 L3 L4 D1 D2 D3 D4	15µH 10% 220µH 10% 180nH 20% 470µH 10% Diodes 1N6263 Schottky 1N6263 Schottky BBY40 varicap 5mm red LED	IC1 X1 SW1 CON1 CON2 PCB	NJ88C33 10.00MHz 5ppm series Miniature slide switch SMC socket SMC socket C33ISS2			
C13 C14 C15 C16	1nF 10% 2p7 ± 0.5pF NPO 10nF 10% 10nF 10%			R12 R13 R14 R15 R16 R17 R18	10 10 22k 2.7k 330R 100 33M	TR1 TR2 TR3	Transistors BFS17 RF NPN BFS17 RF NPN 2N3904 Switching					

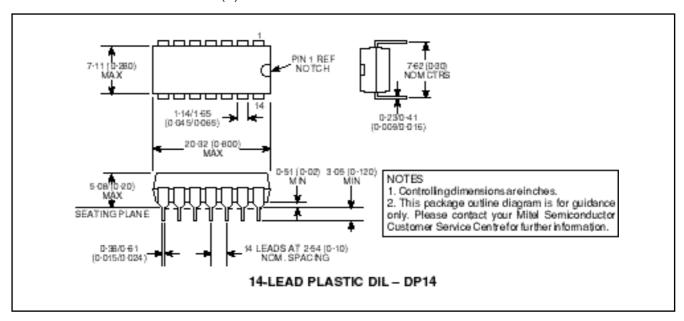
NOTES

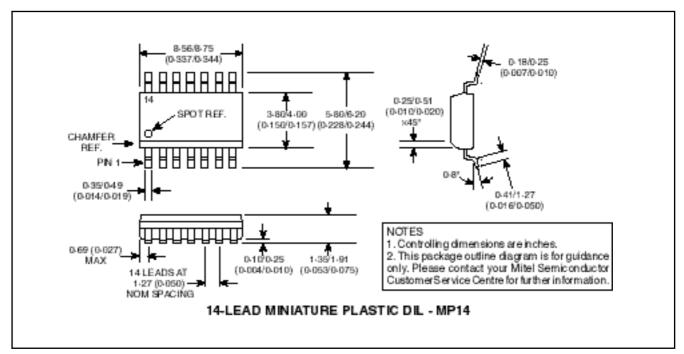
- 1. With the exception of electrolytics, all capacitors are surface mount types.
- 2. All resistors are 0.25W, ±2%.
- 3. C0, C1, C2, C11, C12, C13 and C14 must be low leakage types.
- 4. R18 may be required to optimise VCO close in noise performance.

^{*} Insert C15, delete R4 and R5 if CON2 is to be used to monitor the VCO. Delete C15 insert R4 and R5 if CON is to provide an external source, otherwise short C15 and delete R4, R5 and CON2.

PACKAGE DETAILS

Dimensions are shown thus: mm (in).







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